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The Ohio State University

ADVANCED ADAPTIVE ANTENNA TECHNIQUES

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Quarterly Report 712684-5

Contract N00019-80-C-0181

October 1980

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Naval Air Systems Command  
Washington, D.C. 20361

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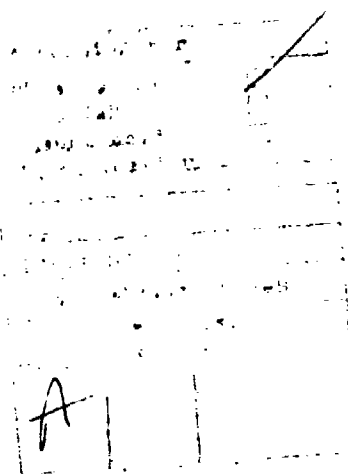
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SECURITY CLASSIFICATION OF THIS PAGE (When Data Entered)

REPORT DOCUMENTATION PAGE		READ INSTRUCTIONS BEFORE COMPLETING FORM
1. REPORT NUMBER	2. GOVT ACCESSION NO.	3. RECIPIENT'S CATALOG NUMBER
	AD-A115155	
4. TITLE (and Subtitle)	5. TYPE OF REPORT & PERIOD COVERED	
Advanced Adaptive Antenna Techniques	Quarterly Report	
7. AUTHOR(s)	6. PERFORMING ORG. REPORT NUMBER	
R. T. Compton, Jr.	ESL 712684-5	
	8. CONTRACT OR GRANT NUMBER(s)	
	Contract N00019-80-C-0181	
9. PERFORMING ORGANIZATION NAME AND ADDRESS		10. PROGRAM ELEMENT, PROJECT, TASK AREA & WORK UNIT NUMBERS
The Ohio State University ElectroScience Laboratory, Department of Electrical Engineering Columbus, Ohio 43212		
11. CONTROLLING OFFICE NAME AND ADDRESS		12. REPORT DATE
Naval Air Systems Command Washington, D. C. 20361		October 1980
		13. NUMBER OF PAGES
		9
14. MONITORING AGENCY NAME & ADDRESS (if different from Controlling Office)		15. SECURITY CLASS. (of this report)
		Unclassified
		15a. DECLASSIFICATION/DOWNGRADING SCHEDULE
16. DISTRIBUTION STATEMENT (of this Report)		
APPROVED FOR PUBLIC RELEASE DISTRIBUTION UNLIMITED		
17. DISTRIBUTION STATEMENT (of the abstract entered in Block 20, if different from Report)		
18. SUPPLEMENTARY NOTES		
19. KEY WORDS (Continue on reverse side if necessary and identify by block number)		
Adaptive Arrays Antennas Interference Communications		
20. ABSTRACT (Continue on reverse side if necessary and identify by block number)		
This report describes progress under Naval Air Systems Command Contract N00019-80-C-0181 during the second quarterly period. Research on advanced adaptive antenna techniques is summarized.		

## CONTENTS

	Page
I. INTRODUCTION	1
II. PROGRESS	1
1. The Effects of Signal Polarization on Array Performance	1
2. Monograph	8
REFERENCES	9



## I. INTRODUCTION

This report describes progress under Naval Air Systems Command Contract N00019-80-C-0181 during the second quarterly period. This contract involves adaptive array studies in two areas: (1) the effects of element patterns and signal polarization on adaptive array performance, and (2) the capability of pulsed and swept CW jamming against adaptive arrays. In addition, a monograph on adaptive arrays is being prepared under this contract.

During the second quarterly period, we have concentrated on two of the above areas: the effects of signal polarization, and the monograph. Progress in these two areas is described below.

## II. PROGRESS

### 1. The Effects of Signal Polarization on Array Performance

During the previous quarter we computed the performance of an array of three tripoles against an arbitrarily but completely polarized\* interference signal [1]. It was found that this array is extremely effective against such interference. The interference can significantly reduce the array output signal-to-interference-plus-noise ratio (SINR) only if it arrives from the same direction and has the same polarization as the desired signal. Therefore, to be effective against such an array, a jammer will be forced to use cross-polarized jamming, i.e., to transmit two

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\*A completely polarized electromagnetic wave is one with a single, fixed polarization. This polarization differs from a partially polarized, or a randomly polarized wave, for which the state of polarization changes randomly with time [2].

independent jamming signals on cross-polarized elements. (This results in a randomly polarized wave, in contrast to the completely polarized wave described above.) During this quarter, we have therefore extended last quarter's work by computing the performance of a single tripole and a 3-tripole array against such cross-polarized jamming.

First, we calculated the output SINR from a single tripole array with a desired signal and a cross-polarized interference signal incident on the array. The SINR was computed as a function of the incidence angles of the desired signal and interference, and of the desired signal polarization. In this analysis, each polarization component of the interference is assumed to have a flat bandlimited power spectrum.

The results of this study show, as expected, that the performance of the single tripole against cross-polarized interference is poorer than it is against a completely polarized interference signal. The reason is that a cross-polarized interference signal uses up two degrees of freedom, whereas a completely polarized interference signal uses up only one. Hence, the tripole has less flexibility left for optimizing desired signal polarization. If the desired signal has linear polarization, the output SINR from a single tripole array with cross-polarized interference is poor for many arrival angles. But if the desired signal is circularly polarized, the SINR is only slightly lower than it is with completely polarized interference.

Figures 1-3 illustrate these comments. Figure 1 shows the tripole antenna, with the angles  $\theta$  and  $\phi$  defined. Figure 2 shows the SINR performance of the tripole with completely polarized interference, and Figure 3 shows the SINR with cross-polarized interference. In both cases, the desired signal has circular polarization.

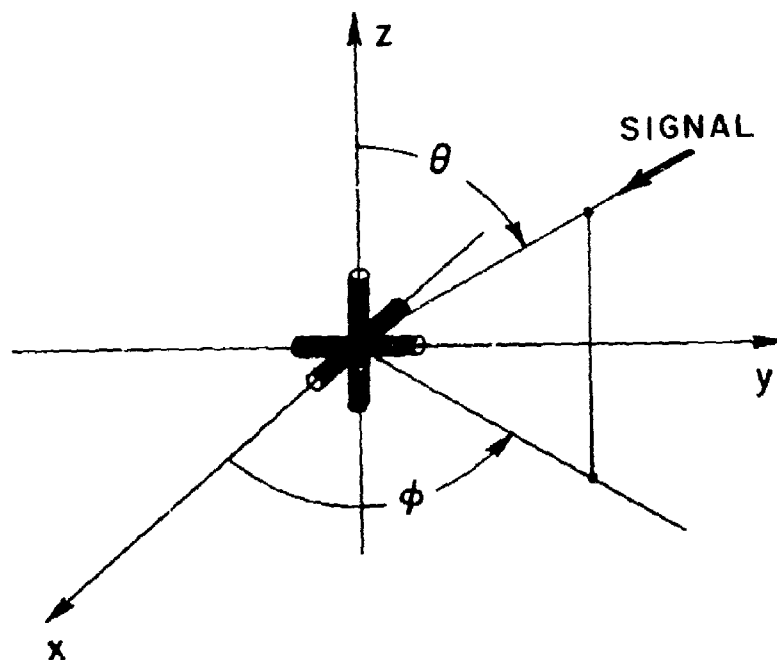


Figure 1. The tripole antenna.

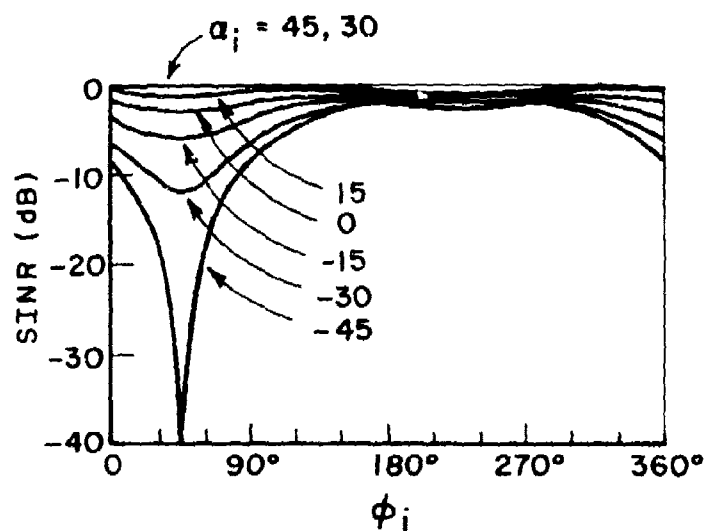


Figure 2. SINR of tripole with completely polarized interference.  
 $(\theta_d=45, \phi_d=45, \alpha_d=45, \beta_d=0, \theta_i=45, \beta_i=0)$   
 $\text{SNR}=0 \text{ dB}, \text{INR}=40 \text{ dB})$

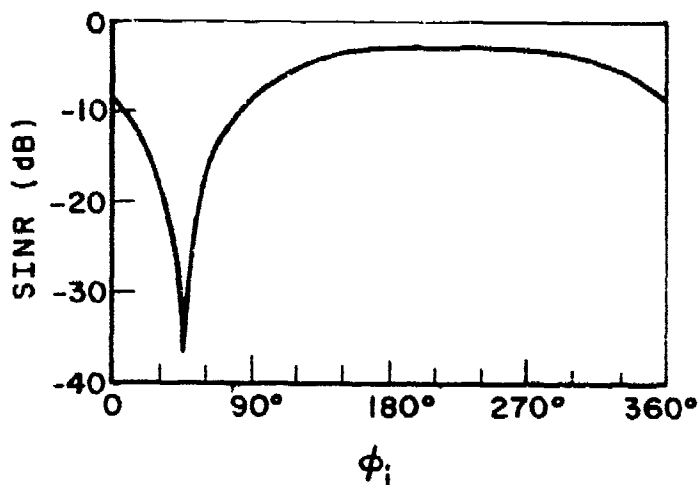


Figure 3. SINR of tripole with cross-polarized interference.  
 $(\theta_d=45^\circ, \phi_d=45^\circ, \alpha_d=45^\circ, \beta_d=0^\circ, \theta_i=45^\circ)$   
 $(\text{SNR} = 0 \text{ dB}, \text{Total INR} = 40 \text{ dB})$

In Figure 2, the desired signal arrives from  $\theta_d = 45^\circ$  and  $\phi_d = 45^\circ$  with polarization ellipticity angle  $\alpha_d = 45^\circ$  and orientation angle  $\beta_d = 0^\circ$ . (These values of  $\alpha_d$  and  $\beta_d$  correspond to left circular polarization.) The desired signal has 0 dB signal-to-noise ratio (SNR). The interference arrives from  $\theta_i = 45^\circ$  with orientation angle  $\beta_i = 0^\circ$  and an interference-to-noise ratio (INR) of 40 dB. Figure 2 shows the output SINR from the array as a function of the interference angle  $\phi_i$ . Seven curves are shown for different values of interference ellipticity:  $\alpha_i = -45^\circ, -30^\circ, -15^\circ, 0^\circ, 15^\circ, 30^\circ$  and  $45^\circ$ . The curve with the lowest SINR is for  $\alpha_i = 45^\circ$ , when the interference has the same polarization as the desired signal. For this case the SINR drops to -40 dB when the interference arrives from the same direction as the desired signal. But for other values of  $\alpha_i$ , the SINR is better, even when the interference arrives from the same direction as the desired signal.



Now consider Figure 3, which shows the output SINR from the same array when the interference is cross-polarized, i.e., when it consists of two independent signals transmitted from the same direction on orthogonal linear polarizations (one polarized in the  $\hat{\theta}$ -direction and the other in the  $\hat{\phi}$ -direction). The desired signal parameters are the same in Figure 3 as in Figure 2. Each component of the interference has an INR of 37 dB, so the total interference power on both polarizations is 40 dB above noise, the same as in Figure 2. Figure 3 again shows the output SINR from the array as a function of  $\phi_i$ .

Comparing Figures 2 and 3 shows that the SINR of the tripole is slightly lower with the cross-polarized jammer than with the completely polarized jammer. However, the difference is small.

An important point to note is that a cross-polarized jammer always forces the SINR to be low when the interference arrives from the same direction as the desired signal. A completely polarized jammer does not necessarily force the SINR to be low when the two signals come from the same direction. Only if the interference polarization is the same as that of the desired signal does it do so. The cross-polarized jammer forces the array to produce a null for both polarizations in the interference direction. Hence the array cannot receive any desired signal from this direction.

During this quarter, we have also computed the performance of an array of three tripoles (i.e., nine elements) against a cross-polarized jammer. These studies are similar to those described above for the single tripole, but also include the effects of interference bandwidth. (Bandwidth has no effect on the performance of a single tripole.) Figures 4 and 5 show typical results. For both figures, the desired signal arrives from  $\theta_d = 45^\circ$  and  $\phi_d = 45^\circ$  with  $\alpha_d = 45^\circ$  and  $\beta_d = 0^\circ$ , i.e., with left circular polarization. The SNR is 0 dB. In both figures, the interference arrives from  $\theta_i = 45^\circ$  with an INR of 37 dB on each polarization component,

or a total INR of 40 dB. The figures again show the array output SINR versus  $\phi_i$ . Figures 4 and 5 differ only in the choice of interference bandwidth. Figure 4 is for zero bandwidth, and Figure 5 is for 10% bandwidth. It is seen that there is relatively little change between the two curves. This is an important point, which we discuss below.

We have learned that an extremely helpful advantage of using co-located cross-polarized elements in an adaptive array is that it greatly reduces the problem of interference bandwidth. Normally, when only one polarization is received by an adaptive array, the SINR from the array drops quickly as interference bandwidth is increased. Even a small bandwidth can produce a substantial drop in SINR. This degradation occurs because nonzero bandwidth causes the interference signals in the different array elements to be partially decorrelated. Hence, with nonzero bandwidth, the interference cannot be nulled as well by subtracting one element signal from another. However, when two or more elements are located at the same position, as in the tripole antenna, there is no interelement time delay for the interference, regardless of its arrival angle. As a result, no decorrelation occurs, and the performance of the tripole is unaffected by interference bandwidth.

In addition, when two or more tripoles are used in a larger array, the SINR performance of the entire array is always at least as good as that of one of the tripoles by itself. (I.e., the array feedback can always turn off the weights in all the elements of the array except for one tripole.) Thus, with wideband interference, the SINR from an array of tripoles cannot be worse than the SINR from a single tripole, and the SINR from a single tripole is not affected by bandwidth. Moreover, if the desired signal has circular polarization, the single tripole SINR is quite respectable, as may be seen in Figure 3. Thus, the use of co-located, cross-polarized elements in an array turns out to have the important side benefit of reducing the problem of interference bandwidth.

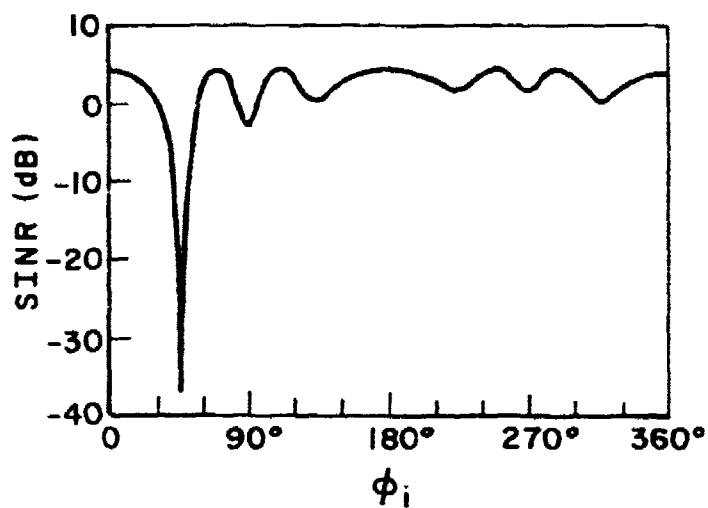


Figure 4. SINR of 3-tripole array with cross-polarized interference (zero bandwidth).  
 $(\theta_d=45, \phi_d=45, \alpha_d=45, \beta_d=0, L/\lambda=2$   
 $SNR=0, \text{ Total INR}=40 \text{ dB})$

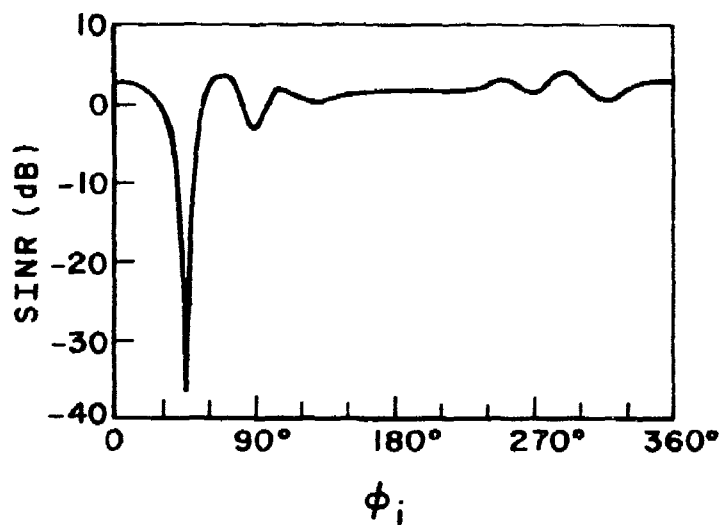


Figure 5. SINR of 3-tripole array with cross-polarized interference (10% bandwidth).  
 $(\theta_d=45, \phi_d=45, \alpha_d=45, \beta_d=0, L/\lambda=2$   
 $SNR=0 \text{ dB}, \text{ Total INR}=40 \text{ dB.})$

Comparing Figures 4 and 5 shows that the performance degradation with 10% bandwidth is quite small. It may be shown that the performance of an adaptive array receiving only one polarization is much poorer with 10% bandwidth. (See, for example, Reference 3 ).

## 2. Monograph

During the second quarter, we have also put considerable effort into the adaptive array monograph. At this date, Chapter III of the monograph is essentially finished. (The first three chapters constitute most of the monograph. Two additional very short chapters are planned on applications.) Presently, 438 types pages of manuscript have been completed.

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